

**SCHOOL OF ELECTRICAL & ELECTRONIC ENGINEERING****ADVANCED DIPLOMA IN ENGINEERING (ELECTRICAL & ELECTRONICS)****EEE606 – CIRCUITS & SIGNALS****Wednesday 30<sup>th</sup> May, 2013    0900 - 1210 hours    Venue: C301****INSTRUCTIONS TO CANDIDATES**

1. Candidates are reminded that they should have no books, notes, paper or other material in their possession unless their use is specifically permitted by "Instructions to Candidates" set out below.
2. Reading time is of 10 minutes duration.
3. Examination time is of 3 hours duration.
4. Write your candidate number at the top of each attached sheet.
5. This paper consists of two (2) sections: A and B.
6. Attempt ALL four (4) questions in Section A and one (1) out of four (4) questions in Section B.
7. Each question carries 20 marks.
8. The following sets of Table are provided: Laplace Transform; Fourier Transform, Standard Integrals, and Trigonometric Identities.
9. Matlab is also available for use.
10. Cellphones are not allowed inside the examination venue.

**SECTION A**

This section is compulsory and all questions are to be attempted.

**QUESTION 1: THEVENIN THEOREM; NORTON THEOREM; SUPERPOSITION THEOREM**

**Q 1 - 1: Thevenin Theorem**

Consider the frequency domain network in Figure 1.

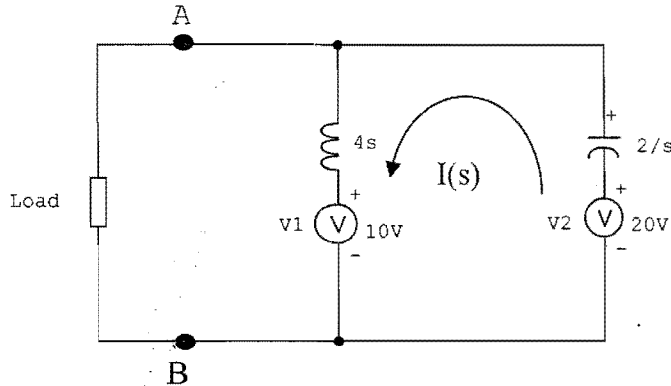


Figure 1: Frequency Domain Network

- (a) Condense the network to the right of terminals A and B, to its Thevenin equivalent circuit, thus obtaining  $V_{Th}$  and  $Z_{Th}$ . [6 marks]
- (b) Determine the short-circuit current when A is connected to B. [3 marks]

**Q1 - 2: Norton Theorem**  
Refer to Figure 1.

- (a) Reduce the network to the right of terminals A and B to its Norton equivalent circuit. [4 marks]

**Q1 - 3: Superposition Theorem**

- (a) Refer to Figure 2. Determine  $I_1$ ,  $I_2$ , and  $I_3$  using the Superposition theorem. [7 marks]

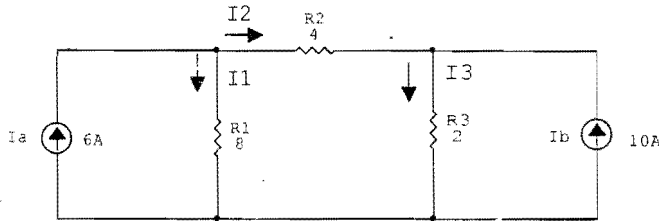


Figure 2

Figure 2

[TOTAL = 20 MARKS]

**QUESTION 2: RLC NETWORK MODELLING & LAPLACE TRANSFORMS APPLICATION**

- Q2 - 1:** Analyze the *RLC* Series circuit shown in Figure 3. The output is taken across the resistor, *R*. Assume zero initial conditions.

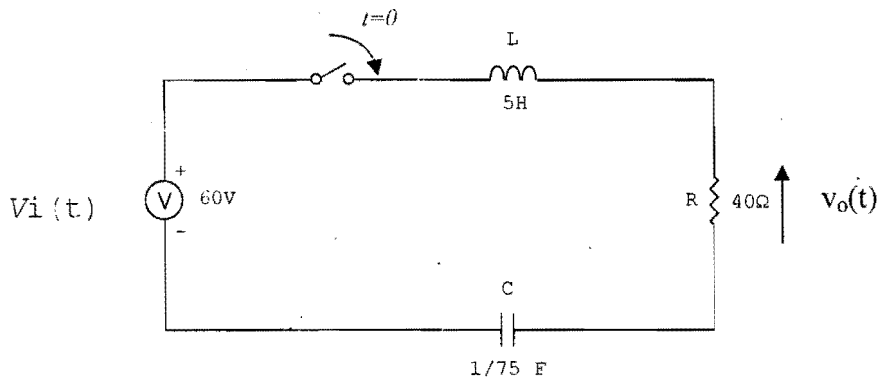


Figure 3

- (a) Derive the Second Order Differential Equation model of the system using Kirchoff's Voltage Law (KVL). [2 marks]
- (b) Transform the Differential Equation from time domain to s-domain using Laplace Transforms, to obtain  $Q(s)$  [5 marks]
- (c) Determine  $q(t)$  from the  $Q(s)$  obtained in (b). [3 marks]

- (d) Express the mathematical model as a first order Differential Equation. [2 marks]
- (e) Determine the Laplace transformed circuit using any suitable method. [3 marks]
- (f) Solve for the Transfer Function,  $G(s)$ , when output is taken across  $R$ . [3 marks]
- (g) Identify, with reason, the type of damping in the system. [2 marks]

[TOTAL = 20 MARKS]

**QUESTION 3: FOURIER SERIES, FOURIER TRANSFORM & CONVOLUTION**

Q3 - 1: A periodic current waveform is given by,  $i(t) = \begin{cases} 20, & 0 \leq t < 2 \\ 0, & 2 \leq t < 4 \end{cases}$  Amps,

and  $i(t+4) = i(t)$ .

(a) Sketch 3 periods of the function  $i(t)$  and label clearly. [2 marks]

(b) Determine the coefficients  $a_0$ ,  $a_n$ , and  $b_n$ . [4 marks]

(c) Synthesize the first 3 terms of the Fourier series of  $v(t)$ . [4 marks]

**Q3 - 2: Fourier Transform & Convolution**

(a) Derive the Fourier Transform of the following systems using the Fourier Transform table.

(i)  $y(t) = 2u(t)\cos 3t$  [2 marks]

(ii)  $q(t) = u(t)e^{-2t} \cos(9t)$  [2 marks]

(b) Using the Convolution Theorem,  $h(t) = (f * g)(t) = \int_0^t f(\tau)g(t-\tau)d\tau$ ,

determine the inverse,  $h(t)$ , of the function,

$$H(s) = \frac{12}{s^2 + 36} \cdot \frac{s}{s^2 + 25} \quad [6 \text{ marks}]$$

[TOTAL = 20 MARKS]

**QUESTION 4: OPERATIONAL AMPLIFIERS, FILTERS, ANALOGUE COMPUTERS**

**Q4 – 1: With the circuit given, determine  $V_o$ ,  $I_1$  and  $I_o$ . [3 + 2 + 2 = 7 marks]**

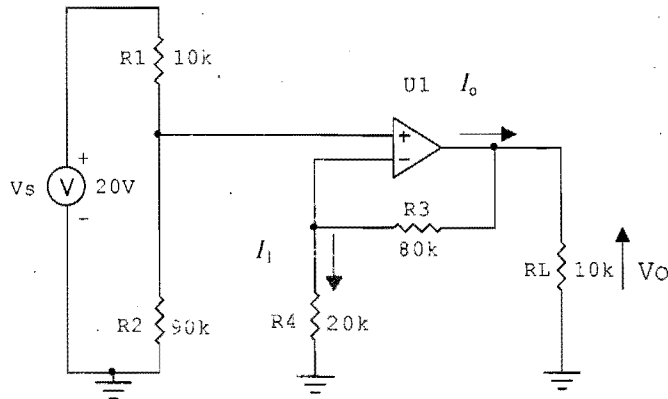


Figure 4

**Q4 - 2: Design a mathematical analogue computer to solve the second order differential equation,  $q'' = -6q' - 3q + 4 \sin 5t$ , using Operational amplifiers. Use Integrators whose time constant  $RC = 1s$ . The initial conditions are  $q'(0) = 0$  and  $q(0) = 1$ . State any assumptions you make and provide a brief explanation on each step of the design. [6 marks]**

**Q4 – 3: The operational amplifier in Figure 6 is configured to control frequency response and acts as a Band Pass Filter. Determine the Transfer Function,  $H(\omega)$ , and identify the Low Pass and High Pass components. [7 marks]**

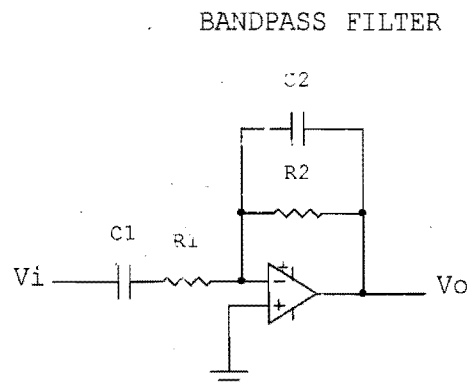


Figure 5: Band Pass Filter

**[TOTAL = 20 MARKS]**

**SECTION B**

Answer only one (1) question in Section B. Please note that this means either Question 5, Question 6, Question 7, or Question 8.

**QUESTION 5: LOOP ANALYSIS & NODAL ANALYSIS**

**Q5 - 1: Loop Analysis**

1. Consider Figure 5, shown. Use Loop (Mesh) analysis to determine currents  $I_1$ ,  $I_2$ ,  $I_3$ , and  $I_0$ . [10 marks]

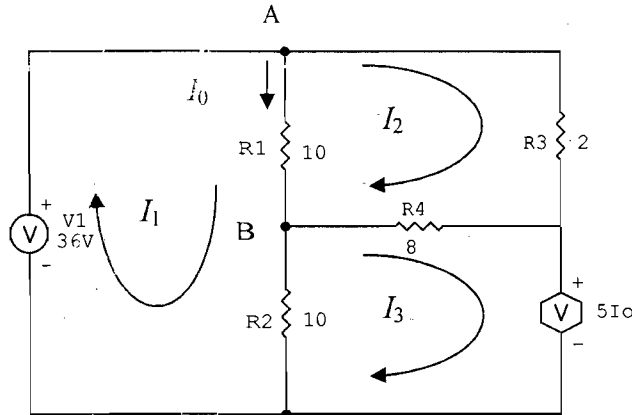


Figure 6

**Q5 - 2: Nodal Analysis**

Consider Figure 6 below.

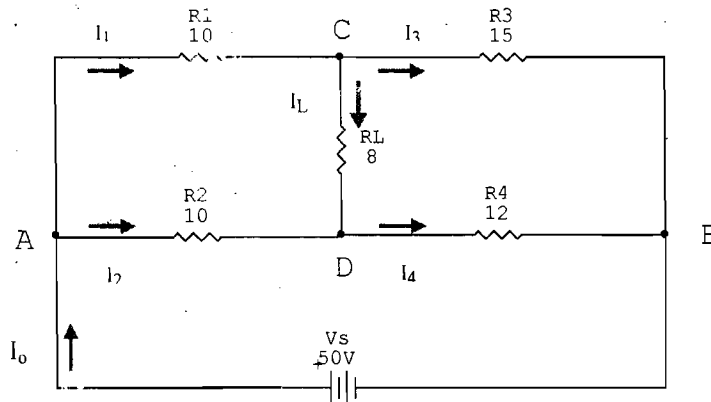


Figure 7

- (a) Use Nodal Analysis to determine  $V_C$  and  $V_D$  [8 marks]
- (b) Determine the load current,  $I_L$ . [2 marks]

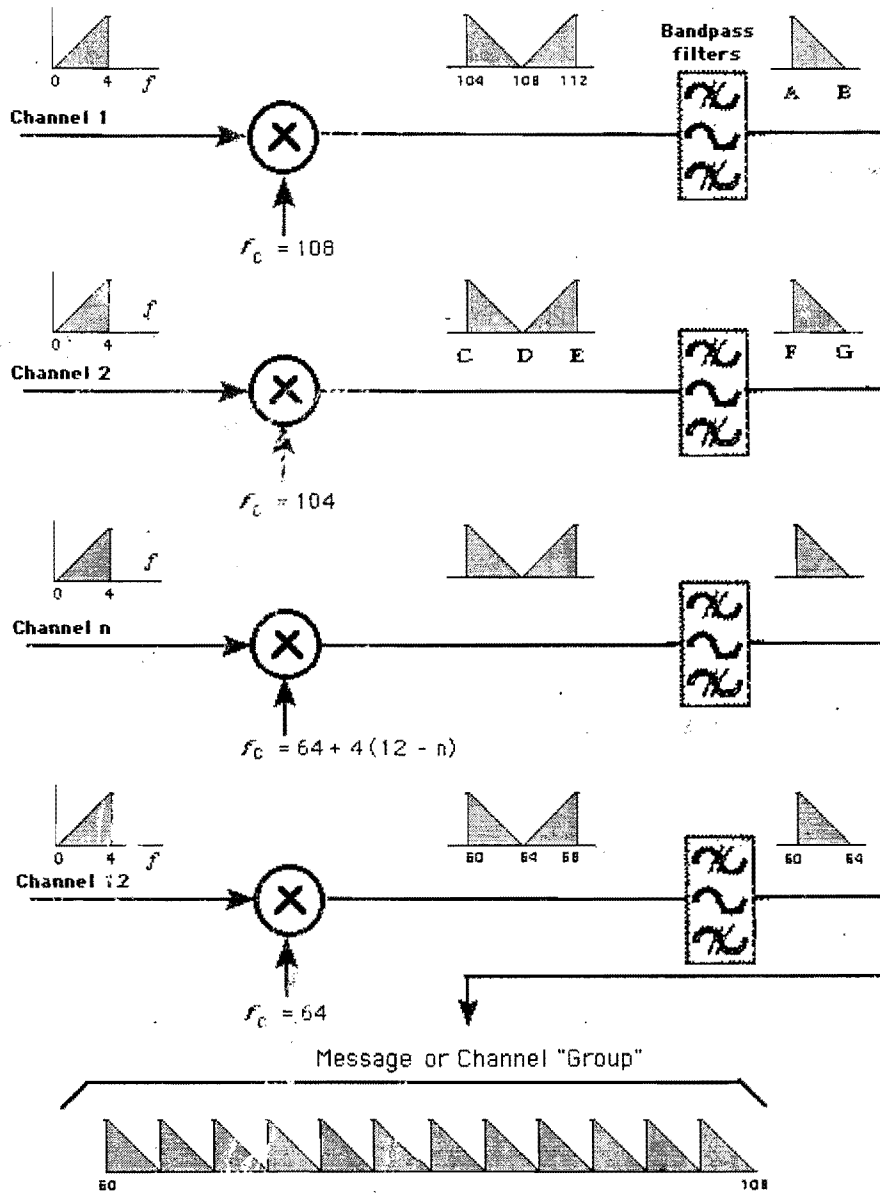
[TOTAL = 20 MARKS]

**QUESTION 6: MULTIPLEXERS, FILTER,**

**Q6-1: Multiplexing**

- (a) Explain the term *Multiplexing* briefly. You may use a diagram to assist you. [2 marks]
- (b) The diagram in Figure 7 shows a schematic of multiple voice channels aggregated into a channel or message group by using Frequency Division Multiplexing (FDM). Complete the diagram by deducing the values for A, B, C, D, E, and F. [4 marks]





Note: All frequencies values are in kHz

Figure 8

- (c) Refer to Table 1, showing the Function Table for the Binary Adder. Design a circuit using the 74LS153 Multiplexer to implement the Sum (S) and the Carry (C). Both outputs S and C are to be present at any time. (Note: Datasheet for 74LS153 is provided).

A	B	K	S	C
0	0	0	0	0
0	0	1	1	0
0	1	0	1	0
0	1	1	0	1
1	0	0	1	0
1	0	1	0	1
1	1	0	0	1
1	1	1	1	1

[8 marks]

Table 1: Binary Adder

- Q6 – 2: The operational amplifier is configured as an Integrator and also a Low Pass Filter in Figure 5. Determine the Transfer Function,  $H(\omega)$ .

[6 marks]

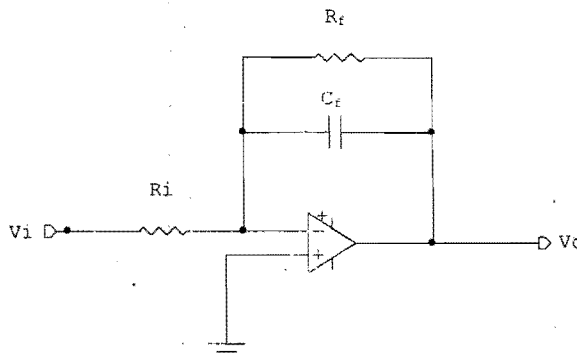


Figure 9: Integrator and Low Pass Filter

**QUESTION 7: SECOND ORDER DIFFERENTIAL EQUATIONS & RLC NETWORK MODELLING APPLICATION**

**Q7-1: Analyze the RLC network shown in Figure 9. The output is taken across the capacitor, C. Assume zero initial conditions.**

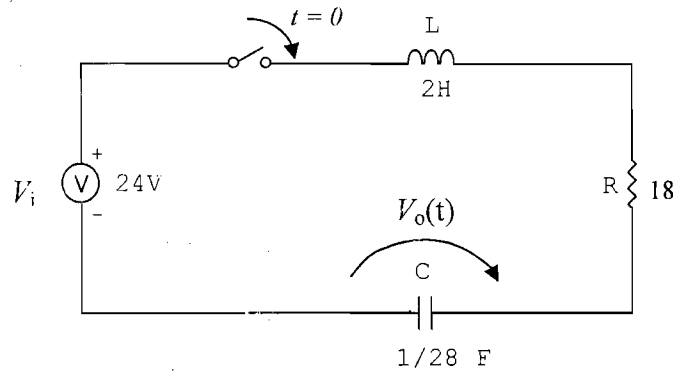


Figure 10

Solve for the following:

- (a) Characteristic (Auxiliary) Equation. [Hint: first find the Second Order Differential equation]. [3 marks]
- (b) Characteristic Roots & Characteristic Modes. [3 marks]
- (c) Complementary Function. [2 marks]
- (d) Briefly explain the term “Zero Input Solution”. [2 marks]
- (e) Particular Integral. [3 marks]
- (f) General Solution. [2 marks]
- (g) Particular Solution. [5 marks]

[TOTAL = 20 MARKS]

**QUESTION 8: COMMUNICATION SYSTEMS & APPLICATION OF FOURIER TRANSFORMS**

**Q8 – 1: From the Fourier Transform, we know that,**

$$F\{\cos 2\pi f_c t\} = \frac{1}{2}[\delta(f - f_c) + \delta(f + f_c)].$$

Applying the Frequency Shifting property produces the Fourier Transform pairs,

$$g(t)e^{j2\pi f_c t} \Leftrightarrow G(f - f_c)$$

The Modulation Property of the Fourier Transform states that,

$$F\{g(t)\cos(2\pi f_c t)\} = F\left\{g(t)\left(\frac{e^{j2\pi f_c t} + e^{-j2\pi f_c t}}{2}\right)\right\} = \frac{1}{2}[G(f - f_c) + G(f + f_c)]$$

The block diagram of the AM Modulator is given in Figure 11 below.

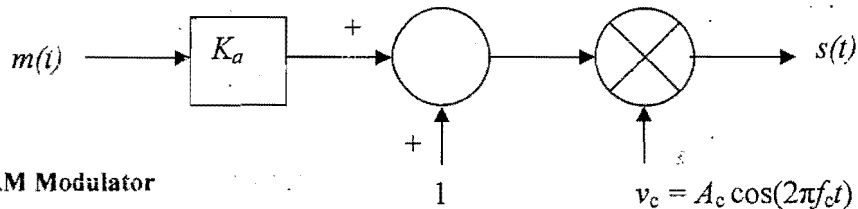


Figure 11: AM Modulator

Supplied are the following facts:

$$m(t) = \cos(2\pi \times 5,000t) V, \quad v_c = 120 \cos(2\pi \times 10^6 t) V$$

- (a) Determine for the expression of the AM signal in the time domain,  $s(t)$ , given the general form  $s(t) = A_c [1 + k_a m(t)] \cos(2\pi f_c t)$ . Assume  $k_a = 1$ .  
[3 marks]

- (b) Synthesize the formula for the AM signal in the Frequency domain,  $S(f)$ , and substitute for the values of the modulating frequency and carrier frequency.  
[7 marks]

Q8 - 2: Given the AM signal,  $s(t) = 250 \left[ 1 + \left( \frac{25}{250} \right) \cos(2\pi \times 10^3 t) \right] \cos(2\pi \times 10^6 t)$  Volts

(i) Sketch the waveform of the AM signal,  $v(t)_{AM}$ . Label it appropriately. [3 marks]

(ii) Derive another equivalent equation of  $s(t)$ . [3 marks]

(iii) Sketch and label the Frequency Spectrum of  $e(t)$  as will be displayed on a Spectrum Analyzer. [2 marks]

(iv) Find the percentage modulation [2 marks]

[TOTAL = 20 marks]

[END OF EXAMINATION]

EEE606-Solution-2013.doc

## TABLE OF LAPLACE TRANSFORMS

$F(s) = L[f(t)]$	$f(t), t \geq 0$
1. 1	$\delta(t_0)$ , Unit Impulse at $t = t_0$
2. $\frac{1}{s}$	1, Unit Step
3. $\frac{n!}{s^{n+1}}$	$t^n$
4. $\frac{1}{s+a}$	$e^{-at}$
5. $\frac{1}{(s+a)^n}$	$\frac{1}{(n-1)!} t^{n-1} e^{-at}$
6. $\frac{a}{s(s+a)}$	$1 - e^{-at}$
7. $\frac{1}{(s+a)(s+b)}$	$\frac{1}{(b-a)} (e^{-at} - e^{-bt})$
8. $\frac{s+\alpha}{(s+a)(s+b)}$	$\frac{1}{(b-a)} [(\alpha-a)e^{-at} - (\alpha-b)e^{-bt}]$
9. $\frac{ab}{s(s+a)(s+b)}$	$1 - \frac{b}{(b-a)} e^{-at} + \frac{a}{(b-a)} e^{-bt}$
10. $\frac{1}{(s+a)(s+b)(s+c)}$	$\frac{e^{-at}}{(b-a)(c-a)} + \frac{e^{-bt}}{(c-a)(a-b)} + \frac{e^{-ct}}{(a-c)(b-c)}$
11. $\frac{s+\alpha}{(s+a)(s+b)(s+c)}$	$\frac{(\alpha-a)e^{-at}}{(b-a)(c-a)} + \frac{(\alpha-b)e^{-bt}}{(c-b)(a-b)} + \frac{(\alpha-c)e^{-ct}}{(a-c)(b-c)}$
12. $\frac{ab(s+\alpha)}{s(s+a)(s+b)}$	$\alpha - \frac{b(\alpha-a)}{(b-a)} e^{-at} + \frac{a(\alpha-a)}{(b-a)} e^{-bt}$
13. $\frac{\omega}{s^2 + \omega^2}$	$\sin \omega t$
14. $\frac{s}{s^2 + \omega^2}$	$\cos \omega t$
15. $\frac{s+a}{s^2 + \omega^2}$	$\frac{\sqrt{\alpha^2 + \omega^2}}{\omega} \sin(\omega t + \phi),$ $\phi = \tan^{-1}\left(\frac{\omega}{\alpha}\right)$
16. $\frac{\omega}{(s+a)^2 + \omega^2}$	$e^{-at} \sin \omega t$

*Please turn over...*

TABLE OF LAPLACE TRANSFORMS

$F(s) = L[f(t)]$	$f(t), t \geq 0$
17. $\frac{(s+a)}{(s+a)^2 + \omega^2}$	$e^{-at} \cos \omega t$
18. $\frac{s+\alpha}{(s+a)^2 + \omega^2}$	$\frac{1}{\omega} [(\alpha-a)^2 + \omega^2]^{\frac{1}{2}} e^{-at} \sin(\omega t + \phi),$ $\phi = \tan^{-1}\left(\frac{\omega}{\alpha-a}\right)$
19. $\frac{\omega_n^2}{s^2 + 2\zeta\omega_n s + \omega_n^2}$	$\frac{\omega_n}{\sqrt{1-\zeta^2}} e^{-\zeta\omega_n t} \sin \omega_n \sqrt{1-\zeta^2} t, \zeta < 1$
20. $\frac{1}{s[(s+a)^2 + \omega^2]}$	$\frac{1}{a^2 + \omega^2} + \frac{1}{\omega\sqrt{a^2 + \omega^2}} e^{-at} \sin(\omega t - \phi),$ $\phi = \tan^{-1}\left(\frac{\omega}{-a}\right)$
21. $\frac{\omega_n^2}{s(s^2 + 2\zeta\omega_n s + \omega_n^2)}$	$1 - \frac{1}{\sqrt{1-\zeta^2}} e^{-\zeta\omega_n t} \sin(\omega_n \sqrt{1-\zeta^2} t + \phi),$ $\phi = \cos^{-1} \zeta, \zeta < 1$
22. $\frac{(s+\alpha)}{s[(s+a)^2 + \omega^2]}$	$\frac{\alpha}{a^2 + \omega^2} + \frac{1}{\omega} \left[ \frac{(\alpha-a)^2 + \omega^2}{a^2 + \omega^2} \right]^{\frac{1}{2}} e^{-at} \sin(\omega t + \phi),$ $\phi = \tan^{-1}\left(\frac{\omega}{\alpha-a}\right) - \tan^{-1}\left(\frac{\omega}{-a}\right)$
23. $\frac{1}{(s+c)[(s+a)^2 + \omega^2]}$	$\frac{e^{-ct}}{(c-a)^2 + \omega^2} + \frac{e^{-at} \sin(\omega t + \phi)}{\omega[(c-a)^2 + \omega^2]^{\frac{1}{2}}},$ $\phi = \tan^{-1}\left(\frac{\omega}{c-a}\right)$

**TABLE OF LAPLACE TRANSFORMS**

24.	$\frac{a}{s^2 - a^2}$	$\sinh at$
25.	$\frac{s}{s^2 - a^2}$	$\cosh at$
26.	$\frac{\omega^2}{s(s^2 + \omega^2)}$	$1 - \cos \omega t$
27.	$\frac{\omega^3}{s^2(s^2 + \omega^2)}$	$\omega t - \sin \omega t$
28.	$\frac{2\omega^3}{(s^2 + \omega^2)^2}$	$\sin \omega t - \omega t \cos \omega t$



TABLE OF LAPLACE TRANSFORMS

$F(s) = L[f(t)]$	$f(t), t \geq 0$
29. $\frac{s}{(s^2 + \omega^2)^2}$	$\frac{t}{2\omega} \sin \omega t$
30. $\frac{s^2}{(s^2 + \omega^2)^2}$	$\frac{1}{2\omega} (\sin \omega t + \omega t \cos \omega t)$
31. $\frac{s}{(s^2 + a^2)(s^2 + b^2)} \quad (a^2 \neq b^2)$	$\frac{1}{b^2 - a^2} (\cos at - \cos bt)$

LAPLACE TRANSFORM OF DERIVATIVES

32.  $sF(s) - f(0)$   $f'(t)$ , First Derivative  
 33.  $s^2 F(s) - sf'(0) - f''(0)$   $f''(t)$ , Second Derivative

GENERAL PROPERTIES OF LAPLACE TRANSFORMS

- Linearity  $L[af_1(t) + bf_2(t)] = aF_1(s) + bF_2(s)$
- Scaling  $L[f(at)] = \frac{1}{a} F\left(\frac{s}{a}\right)$
- Transform of a Derivative  $L[f^{(n)}(t)] = s^n F(s) - s^{n-1} f(0) - s^{n-2} f'(0) - \dots - sf^{(n-2)}(0) - f^{(n-1)}(0)$
- First Shifting Theorem  $L[e^{at} f(t)] = F(s - a)$   
(Complex Translation;  $s$ -shifting)
- Second Shifting Theorem  $L[u(t - a)f(t - a)] = e^{-as} F(s)$   
(Real Translation;  $t$ -shifting)
- Integration  $L\left[\int_0^t f(\tau) d\tau\right] = \frac{1}{s} F(s)$
- Complex Differentiation  $L[tf(t)] = -\frac{dF(s)}{ds}$
- Final Value Theorem  $f(\infty) = \lim_{s \rightarrow 0} sF(s)$
- Initial Value Theorem  $f(0+) = \lim_{s \rightarrow \infty} sF(s)$

## Table of Fourier Transform Pairs

Function, $f(t)$	Fourier Transform, $F(\omega)$
<i>Definition of Inverse Fourier Transform</i> $f(t) = \frac{1}{2\pi} \int_{-\infty}^{\infty} F(\omega) e^{j\omega t} d\omega$	<i>Definition of Fourier Transform</i> $F(\omega) = \int_{-\infty}^{\infty} f(t) e^{-j\omega t} dt$
$f(t - t_0)$	$F(\omega) e^{-j\omega t_0}$
$f(t) e^{j\omega_0 t}$	$F(\omega - \omega_0)$
$f(\alpha t)$	$\frac{1}{ \alpha } F\left(\frac{\omega}{\alpha}\right)$
$F(t)$	$2\pi f(-\omega)$
$\frac{d^n f(t)}{dt^n}$	$(j\omega)^n F(\omega)$
$(-jt)^n f(t)$	$\frac{d^n F(\omega)}{d\omega^n}$
$\int_{-\infty}^t f(\tau) d\tau$	$\frac{F(\omega)}{j\omega} + \pi F(0) \delta(\omega)$
$\delta(t)$	1
$e^{j\omega_0 t}$	$2\pi \delta(\omega - \omega_0)$
$\text{sgn}(t)$	$\frac{2}{j\omega}$

$j \frac{1}{\pi}$	$\text{sgn}(\omega)$
$u(t)$	$\pi\delta(\omega) + \frac{1}{j\omega}$
$\sum_{n=-\infty}^{\infty} F_n e^{jn\omega_0 t}$	$2\pi \sum_{n=-\infty}^{\infty} F_n \delta(\omega - n\omega_0)$
$\text{rect}\left(\frac{t}{\tau}\right)$	$\tau \text{Sa}\left(\frac{\omega\tau}{2}\right)$
$\frac{B}{2\pi} \text{Sa}\left(\frac{Bt}{2}\right)$	$\text{rect}\left(\frac{\omega}{B}\right)$
$\text{tri}(t)$	$\text{Sa}^2\left(\frac{\omega}{2}\right)$
$A \cos\left(\frac{\pi}{2\tau}\right) \text{rect}\left(\frac{t}{2\tau}\right)$	$\frac{A\pi}{\tau} \frac{\cos(\omega\tau)}{(\pi/2\tau)^2 - \omega^2}$
$\cos(\omega_0 t)$	$\pi[\delta(\omega - \omega_0) + \delta(\omega + \omega_0)]$
$\sin(\omega_0 t)$	$\frac{\pi}{j} [\delta(\omega - \omega_0) - \delta(\omega + \omega_0)]$
$u(t) \cos(\omega_0 t)$	$\frac{\pi}{2} [\delta(\omega - \omega_0) + \delta(\omega + \omega_0)] + \frac{j\omega}{\omega_0^2 - \omega^2}$
$u(t) \sin(\omega_0 t)$	$\frac{\pi}{2j} [\delta(\omega - \omega_0) - \delta(\omega + \omega_0)] + \frac{\omega^2}{\omega_0^2 - \omega^2}$
$u(t)e^{-\alpha t} \cos(\omega_0 t)$	$\frac{(\alpha + j\omega)}{\omega_0^2 + (\alpha + j\omega)^2}$

$u(t)e^{-\alpha t} \sin(\omega_0 t)$	$\frac{\omega_0}{\omega_0^2 + (\alpha + j\omega)^2}$
$e^{-\alpha t }$	$\frac{2\alpha}{\alpha^2 + \omega^2}$
$e^{-t^2/(2\sigma^2)}$	$\sigma\sqrt{2\pi} e^{-\sigma^2\omega^2/2}$
$u(t)e^{-\alpha t}$	$\frac{1}{\alpha + j\omega}$
$u(t)te^{-\alpha t}$	$\frac{1}{(\alpha + j\omega)^2}$

➤ **Trigonometric Fourier Series**

$$f(t) = a_0 + \sum_{n=1}^{\infty} (a_n \cos(\omega_0 nt) + b_n \sin(\omega_0 nt))$$

where

$$a_0 = \frac{1}{T} \int_0^T f(t) dt, \quad a_n = \frac{2}{T} \int_0^T f(t) \cos(\omega_0 nt) dt, \text{ and}$$

$$b_n = \frac{2}{T} \int_0^T f(t) \sin(\omega_0 nt) dt$$

➤ **Complex Exponential Fourier Series**

$$f(t) = \sum_{n=-\infty}^{\infty} F_n e^{j\omega_0 nt}, \text{ where } F_n = \frac{1}{T} \int_0^T f(t) e^{-j\omega_0 nt} dt$$

### *Some Useful Mathematical Relationships*

$\cos(x) = \frac{e^{jx} + e^{-jx}}{2}$
$\sin(x) = \frac{e^{jx} - e^{-jx}}{2j}$
$\cos(x \pm y) = \cos(x)\cos(y) \mp \sin(x)\sin(y)$
$\sin(x \pm y) = \sin(x)\cos(y) \pm \cos(x)\sin(y)$
$\cos(2x) = \cos^2(x) - \sin^2(x)$
$\sin(2x) = 2\sin(x)\cos(x)$
$2\cos^2(x) = 1 + \cos(2x)$
$2\sin^2(x) = 1 - \cos(2x)$
$\cos^2(x) + \sin^2(x) = 1$
$2\cos(x)\cos(y) = \cos(x-y) + \cos(x+y)$
$2\sin(x)\sin(y) = \cos(x-y) - \cos(x+y)$
$2\sin(x)\cos(y) = \sin(x-y) + \sin(x+y)$

### Useful Integrals

$\int \cos(x) dx$	$\sin(x)$
$\int \sin(x) dx$	$-\cos(x)$
$\int x \cos(x) dx$	$\cos(x) + x \sin(x)$
$\int x \sin(x) dx$	$\sin(x) - x \cos(x)$
$\int x^2 \cos(x) dx$	$2x \cos(x) + (x^2 - 2) \sin(x)$
$\int x^2 \sin(x) dx$	$2x \sin(x) - (x^2 - 2) \cos(x)$
$\int e^{ax} dx$	$\frac{e^{ax}}{a}$
$\int x e^{ax} dx$	$e^{ax} \left[ \frac{x}{a} - \frac{1}{a^2} \right]$
$\int x^2 e^{ax} dx$	$e^{ax} \left[ \frac{x^2}{a} - \frac{2x}{a^2} + \frac{2}{a^3} \right]$
$\int \frac{dx}{\alpha + \beta x}$	$\frac{1}{\beta} \ln \alpha + \beta x $
$\int \frac{dx}{\alpha^2 + \beta^2 x^2}$	$\frac{1}{\alpha\beta} \tan^{-1} \left( \frac{\beta x}{\alpha} \right)$

## DM74LS153 Dual 1-of-4 Line Data Selectors/Multiplexers

### General Description

Each of these data selectors/multiplexers contains inverters and drivers to supply fully complementary, on-chip, binary decoding data selection to the AND-OR-invert gates. Separate strobe inputs are provided for each of the two four-line sections.

### Features

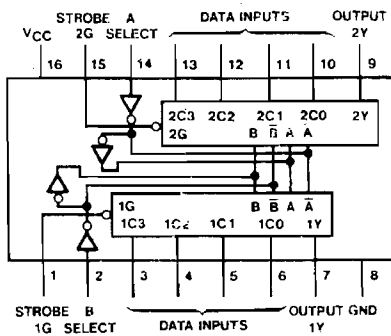
- Permits multiplexing from N lines to 1 line
- Performs at parallel-to-serial conversion
- Strobe (enable) line provided for cascading (N lines to n lines)
- High fan-out, low impedance, totem pole outputs
- Typical average propagation delay times
  - From data 14 ns
  - From strobe 19 ns
  - From select 22 ns
- Typical power dissipation 31 mW

### Ordering Code:

Order Number	Package Number	Package Description
DM74LS153M	M16A	16-Lead Small Outline Integrated Circuit (SOIC), JEDEC MS-012, 0.150 Narrow
DM74LS153N	N16E	16-Lead Plastic Dual-In-Line Package (PDIP), JEDEC MS-001, 0.300 Wide

Devices also available in Tape and Reel. Specify by appending the suffix letter "X" to the ordering code.

### Connection Diagram



### Function Table

Select Inputs		Data Inputs				Strobe	Output
B	A	C0	C1	C2	C3	G	Y
X	X	X	X	X	X	H	L
L	L	L	X	X	X	L	L
L	L	H	X	X	X	L	H
L	H	X	L	X	X	L	L
L	H	X	H	X	X	L	H
H	L	X	X	L	X	L	L
H	L	X	X	H	X	L	H
H	H	X	X	X	L	L	L
H	H	X	X	X	H	L	H

Select inputs A and B are common to both sections.  
H = HIGH Level  
L = LOW Level  
X = Don't Care

DM74LS153

### Logic Diagram

